An Efficient Design of Parallel Pipelined FFT Architecture

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Abstract: This paper presents a new parallel pipelined architecture to compute Discrete Fourier Transform (DFT) using FFT architecture. This particular architecture uses folding transformation technique as well as register minimization technique for the design of FFT architecture. Novel FFT architectures for the computation of complex and real valued signals are derived. Pipelining is used to reduce the power consumption. Parallel processing and pipelining exploits concurrency. Parallel processing also aids to the reduction of power consumption by reducing the supply voltage. The power consumption is reduced very effectively using the parallel architecture. This paper also includes various techniques to reduce the computation time and power using different types of multipliers.

Keywords: Fast Fourier Transform, Folding transformation, Register Minimization, Pipelining.

1. Introduction

Fast Fourier Transform (FFT) is a commonly used technique for the computation of Discrete Fourier Transform (DFT). DFT computations are required in the fields like filtering, spectral analysis etc. to calculate the frequency spectrum or to identify a system's frequency response from its impulse response and vice versa. FFT is used in digital video broadcasting and OFDM systems. Much research has been carried out to design pipelined architectures for computation of FFT. The basic one is Radix-2 FFT. Based on the radix-2 FFT approach many algorithms have been developed which includes radix-4 [4], split-radix [3], radix- [5] etc. A popularly known algorithm is Cooley-Tukey radix-2 FFT [2].

Radix-2 Multipath Delay Commutator (R2MDC) [6] is a classical approach for pipelined implementation of FFT architecture. Radix-2 Single-path Delay Feedback (R2SDF) [7] is another approach with reduced memory obtained by a standard usage of storage buffer in R2MDC. Most of the algorithms require hardware complexity and there is no complete hardware utilisation. The basic aspects like high throughput and low power consumption are required to speed and power requirements keeping the hardware overhead to a minimum. This paper presents a technique to design the architecture from FFT flow graph. Folding transformation [8],[9] and register minimization [8],[10],[11]are the two important steps included in this FFT algorithm.

Folding Transformation is a technique in which many butterflies in the same column can be mapped into one butterfly

unit. If we consider an FFT of size N, then 2-parallel architecture can be obtained if we consider the folding factor to be N/2 or 4-parallel architecture if considering a folding factor of N/4. By selecting the appropriate folding sets we can derive the FFT architectures. The folding sets are designed in a way to reduce the number of storage elements and also the latency. The prior FFT architectures had no systematic way of approach. This architecture simplifies the design of FFT and is a systematic approach towards the design of FFT with arbitrary level of parallelism. These are derived either in Decimation-In-Time (DIT) or Decimation-In-Frequency (DIF) flow graphs. FFT architectures can be derived for different radices.

Parallel pipelined architectures for the computation of real valued signals (RFFT) based on radix-2² and radix-2³ and different architectures for the computation of complex valued signals (CFFT) are carried out earlier. This paper is organised into VI sections where section II represents the folding transformation and register minimization based FFT architecture, section III and IV explains about the proposed architecture for CFFT and RFFT respectively. In section V a comparative study is conducted using different multipliers to identify the multiplier which makes use of minimum number of clock cycles in FFT computation. Finally, section VI certain conclusions are drawn from the comparative study.

2. FFT Architecture via Folding Transformation

Folding Transformation and Register minimization techniques are used to derive several FFT architectures. The whole process is explained with the help of 8-point radix-2 DIF FFT which can be extended to different radices. The flow graph of 8-point radix-2 DIF FFT is illustrated in Fig. 1. The graph has three stages and each stage consists of a set of butterflies and multipliers. The Data Flow Graph of Fig.1 is shown in Fig. 2 where each node represents a computation.



Figure 1: Flow graph of a radix-2 8-point DIF FFT



Figure 2: Data Flow graph of a radix-2 8-point DIF FFT

DFG is subjected to folding transformation in order to derive a pipelined architecture. For this we require a folding set, which is a set of operations executed by the same functional unit.

Every folding set contains number of entries which are called the folding factors. A folding set may include null operations also. Consider two nodes represented as U and V which are connected by an edge e with w(e) delays. The *l*-th iteration of these nodes be scheduled at Kl+u and Kl+v where is the number of entries and *u* and *v* are the folding orders. The folding equation is represented as

$$D_F(U \rightarrow V) = Kw(e) - P_U + v - u \qquad (1)$$

where P_U is the number of pipeline stages. For the DFG in Fig. 2 consider the folding set shown below.

$$\begin{split} &A = \{\phi, \phi, \phi, \phi, A0, A1, A2, A3\} \\ &B = \{B2, B3, \phi, \phi, \phi, \phi, B0, B1\} \\ &C = \{C1, C2, C3, \phi, \phi, \phi, \phi, C0\}. \end{split}$$

We assume that the butterfly operations do not have any pipeline stages. Prior to deriving the folded architecture the folded equations in (1) are to be written for all the edges as shown in (2). $D_F (A_0 \rightarrow B_0) = 2$ means there is an edge with weight 2 from node A to B in the folded DFG. After obtaining the folding equations as shown below, we have to determine whether the folding sets are feasible or not.

$D_F(A0 \to B0) = 2$	$D_F(B0 \to C0) = 1$	
$D_F(A0 \to B2) = -4$	$D_F(B0 \to C1) = -6$	
$D_F(A1 \to B1) = 2$	$D_F(B1 \to C0) = 0$	
$D_F(A1 \to B1) = -4$	$D_F(B1 \to C1) = -7$	
$D_F(A2 \to B0) = 0$	$D_F(B2 \to C2) = 1$	
$D_F(A2 \rightarrow B2) = -6$	$D_F(B2 \to C3) = 2$	
$D_F(A3 \to B1) = 0$	$D_F(B3 \to C2) = 0$	
$D_F(A3 \to B3) = -6$	$D_F(B3 \to C3) = 1.$	
		(2)

In the equations obtained some negative delays are observed which needs to be removed. To make sure that the folded architecture has non-negative number of delay the DFG can be pipelined as shown in Fig. 3. The folding equations for the pipelined DFG are given by





Figure 3: Pipelined Data Flow graph of a radix-2 8-point DIF FFT

From the above equations we can see 24 registers are required for implementing the folded architecture. As a next step a technique called Lifetime analysis [8],[10],[11] is employed to design the architecture with the minimum number of delays. A lifetime chart is obtained as shown in Fig. 4 for one stage of the 8-point DFG.



Figure 4: Lifetime chart for variables y0, y1, ..., y_n

From the lifetime chart we can analyse that we require only 4 registers to implement the design while considering the outputs of nodes A0, A1, A2 and A3 in the DFG instead of the 16 registers which was used in the straight forward

implementation. Next step is Register allocation as shown in Fig. 5.

	I/P	R1	R2	R3	R4
4	y0,y4				
5	y1,y5	y4		+y0 \	
6	(y2)y6	y5	y4	→y1	60
7	(y3)y7	у6	y5	y4	V 1
8		y7	(y6)	y5)
9			۲		¥5

Figure 5: Register allocation table for data shown in figure 4.

From the folding equations and the table in Fig. 5 the architecture in Fig. 6 can be derived. We can see from the folding sets that half of the time null operations are being executed and therefore the hardware utilization is only 50%.



Figure 6:Folded architecture for the DFG in Figure 3

3. Power Consumption

A comparison is made on the basis of power between serial and parallel FFT architectures. The dynamic power consumption of a CMOS circuit is obtained using the equation shown below.

$$P_{\rm ser} = C_{\rm ser} V^2 f_{\rm ser}$$

Where C_{ser} is the total capacitance of a serial circuit, V is the supply voltage and f_{ser} is the clock frequency. P_{ser} is the power consumption of the serial architecture. For an L-parallel system the clock frequency is f_{ser}/L . So the power consumption in an L-parallel system is represented as follows.

$$P_{\rm par} = C_{\rm par} V^2 \frac{f_{\rm ser}}{L}$$

where C_{par} is the total capacitance of the *L*-parallel system.

4. Comparison of 8-point FFT Architectures using different Multipliers.

A comparison is made using different Multipliers in FFT architecture and simulated using ModelSim 6.5e. The 8-point FFT architecture is simulated thrice, each time with one of the three multipliers namely Vedic Multiplier, Array Multiplier and Baugh Wooley Multiplier. A comparison table is obtained for 8-point pipelined parallel FFT architecture using the three different multipliers regarding the time of operation.

4.1 Vedic Multiplier

Vedic Mathematics is an Indian mathematics technique based on 16 sutras. It is a high speed complex multiplier.



4.2 Array Multiplier

Array Multiplier has a regular structure. It is based on add and shift algorithm.



4.3 Baugh Wooley Multiplier

It is used for both signed and unsigned multiplication. It operates on signed operands with 2's complement representation. It uses only fewer steps and lesser adders.

The table below shows the result of comparison of 8-point FFT architectures using different multipliers.

Table 1: 8-	point FFT	'Architecture	using	different	Multipliers
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Sl. No	Multiplier	Time of operation
1	Vedic multiplier	6500ns
2	Array multiplier	6900ns
3	Baugh Wooley multiplier	8600ns

From the table we can understand that Vedic multiplier is very much efficient in terms of speed of operation. Based on the above observation architectures of 16-point FFT are designed for both complex and real inputs using Vedic multiplier.

5. Architecture with Complex Inputs (CFFT)

This section presents parallel architecture for complex valued signals based on radix-2 and radix- 2^3 algorithms. The approach presented in the previous section can be used to derive all these architectures.

5.1 2-parallel radix-2 FFT Architecture

Fig. 7 shows the DFG of radix-2 DIF FFT for N=16 where all the nodes represent radix-2 butterfly operations.



Figure 7: DFG of a radix-2 16-point pipelined DIF FFT Consider the folding sets

$$\begin{split} &A = \{A0, A2, A4, A6, A1, A3, A5, A7\} \\ &B = \{B5, B7, B0, B2, B4, B6, B1, B3\}, \\ &C = \{C3, C5, C7, C0, C2, C4, C6, C1\} \\ &D = \{D2, D4, D6, D1, D3, D5, D7, D0\}. \end{split}$$

We can observe that here the folding sets does not contain any null operations. Thus we can derive the folded architecture using the steps used in the previous section. In this architecture two input samples are processed at the same time. The hardware utilization is 100%. The architecture is shown in Fig. 8.



Figure 8: 2-parallel architecture of a redix-2 16-point DIF complex FFT

In a similar way the 2-parallel architecture for radix-2 DIT FFT can also be derived using the folding set as follows. Fig. 9 represents the pipelined DFG and fig. 10 shows the 2-parallel architecture.



Figure 9: DFG of a radix-2 16-point pipelined DIT FFT



Figure 10: 2-parallel architecture of a redix-2 16-point DIT complex FFT

5.1 4-Parallel Radix-2 FFT Architecture

Using the algorithm used in the previous section we can obtain the pipelined DFG as in Fig. 11. Consider the folding set shown below using which 4-parallel architecture can be derived



Figure 11: DFG of a radix-2 16-point pipelined DIF FFT

(C'3)

D'3 X7



Figure 12: 4-parallel Architecture of a Radix-2 16-point DIF complex FFT

6. Architecture with Real Inputs (RFFT)

The input sequence x[n] for RFFT is considered to be real. If x[n] is real then output X[k] is symmetric.

ie ; $X[N-k] = X^{*}[k]$.

Using this property (N/2) - 10utputs can be removed which are redundant. A new approach in identifying these redundant samples is proposed in [12]. The shaded regions of the Fig. 13 can be removed as they are all redundant samples identified using the approach in [12] and only N/2 + 1 outputs of the FFT are required.

6.1 2-Parallel Radix-2 Architecture

The DFG of this architecture is same as Fig. 7 and the folding set is as follows.

$$\begin{split} &A = \{A0, A2, A4, A6, A1, A3, A5, A7\} \\ &B = \{B5, B7, B0, B2, B4, B6, B1, B3\}, \\ &C = \{C3, C5, \phi, C0, C2, C4, \phi, C1\} \\ &D = \{D2, D4, \phi, D1, \phi, D5, \phi, D0\}. \end{split}$$

The architecture is similar to that shown in Fig. 8 except that first two stages will contain a real data path. The hardware complexity is same as that of the CFFT.

6.2 2-parallel Radix-2² Architecture

Two different scheduling approaches are used to derive two different architectures. It is mainly done by changing the folding order of the butterfly nodes.

1) *Scheduling Method 1:* The parallel-pipelined architecture is shown in Fig. 14 obtained from Fig. 13. The folding set used is

$$\begin{split} &A = \{A0, A2, A4, A6, A1, A3, A5, A7\} \\ &B = \{B5, B7, B0, B2, B4, B6, B1, B3\}, \\ &C = \{C3, C5, \phi, C0, C2, C4, \phi, C1\} \\ &D = \{D2, D4, \phi, D1, \phi, D5, \phi, D0\}. \end{split}$$



Figure 13: Flow Graph of a radix-2² 16-point pipelined DIF FFT.



Figure 14: 2-parallel architecture of a redix-2² 16-point DIF RFFT

The scheduling for the architecture is shown in Fig. 15.



Figure 15: simplified flow graph with scheduling 1.

2) Scheduling Method 2: This method reduces the number of delay elements and slightly modifies the architecture [1], [8]. The folding set is as follows.

$$\begin{split} &A = \{A0, A1, A2, A3, A4, A5, A6, A7\} \\ &B = \{B4, B5, B6, B7, B0, B1, B2, B3\}, \\ &C = \{C2, C3, C4, C5, \phi, \phi, C0, C1\} \\ &D = \{D1, D2, \phi, D4, D5, \phi, \phi, D0\}. \end{split}$$

The modified architecture is shown in Fig. 16 and the scheduling is shown in Fig. 17.



Figure 16: 2-parallel architecture of a redix-2² 16-point DIF RFFT



Figure 17: simplified flow graph with scheduling 2

7. Conclusion

This paper presents a pipelined parallel FFT architecture which has a lesser power consumption compared to serial FFT architectures. It also has the advantage of complete hardware utilization. For a high speed Pipelined parallel FFT architecture a Vedic multiplier can be employed in the particular design. Thus an efficient design can be obtained in terms of power and speed.

References

- Pipelined Parallel FFT Architectures via Folding Transformation, Manohar Ayinala, Student Member, IEEE, Michael Brown, and Keshab K. Parhi, Fellow, IEEE transactions on very large scale integration (vlsi) systems, vol. 20, no. 6, June 2012.
- [2] J. W. Cooley and J. Tukey, "An algorithm for machine calculation of complex fourier series," Math. Comput., vol. 19, pp. 297–301, Apr. 1965.
- [3] P. Duhamel, "Implementation of split-radix FFT algorithms for complex, real, and real-symmetric data," *IEEE Trans. Acoust., Speech, Signal Process.*, vol. 34, no. 2, pp. 285–295, Apr. 1986.
- [4] A. V. Oppenheim, R.W. Schafer, and J. R. Buck, Discrete-Time Signal Processing, 2nd ed. Englewood Cliffs, NJ: Prentice-Hall, 1998.
- [5] S. He and M. Torkelson, "A new approach to pipeline FFT processor," in Proc. of IPPS, 1996, pp. 766–770.
- [6] L. R. Rabiner and B. Gold, *Theory and Application of Digital Signal Processing*. Englewood Cliffs, NJ: Prentice-Hall, 1975.
- [7] E. H. Wold and A. M. Despain, "Pipeline and parallelpipeline FFT processors for VLSI implementation," *IEEE Trans. Comput.*, vol. C-33, no. 5, pp. 414–426, May 1984.
- [8] Keshab K Parhi, VLSI Digital Signal Processing Systems: Design and implementation, , Hoboken, NJ: Wiley.1
- [9] K. K. Parhi, "Calculation of minimum number of registers in arbitrary life time chart," *IEEE Trans. Circuits Syst. II, Exp. Briefs*, vol. 41, no. 6, pp. 434–436, Jun. 1995.

- [10] K. K. Parhi, C. Y. Wang, and A. P. Brown, "Synthesis of control circuits in folded pipelined DSP architectures," IEEE J. Solid-State Circuits, vol. 27, no. 1, pp. 29–43, Jan. 1992.
- [11] K. K. Parhi, "Systematic synthesis of DSP data format converters using lifetime analysis and forward-backward register allocation," *IEEE Trans. Circuits Syst. II, Exp. Briefs*, vol. 39, no. 7, pp. 423–440, Jul. 1992.
- [12] M. Garrido, K. K. Parhi, and J. Grajal, "A pipelined FFT architecture for real-valued signals," IEEE Trans. Circuits Syst. I, Reg. Papers, vol. 56, no. 12, pp. 2634–2643, Dec. 2009.